

ECE3084-L19 Parseval's Theorem

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Parseval's Theorem:

$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(j\omega)|^2 d\omega$$


Note: The change of variables $\omega = 2\pi f$ would make the $1/2\pi$ go away.

Energy in time domain = Energy in frequency domain

IF: $x(t)$ and $X(j\omega)$ are "square integrable" (both integrals must converge to finite numbers).

→ This theorem does not apply to things like δ fns & sinusoids.

View δ fn as the limit of an area 1 rect as its width goes to zero:


$$\int_0^1 1^2 dt = 1 \quad \int_0^{1/2} 2^2 dt = \frac{4}{2} = 2 \quad \int_0^{1/h} h^2 dt = \frac{h^2}{h} = h \quad \int_{0^-}^{0^+} \delta^2(t) dt \stackrel{?}{=} \infty$$

Version for periodic functions, where a_k are the FS coeffs of $x(t)$:

$$\frac{1}{T_0} \int_0^{T_0} |x(t)|^2 dt = \sum_{k=-\infty}^{\infty} |a_k|^2$$

IF both sides are finite.

Generalized form:

$$\int_{-\infty}^{\infty} F(t) g^*(t) dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\omega) G^*(j\omega) d\omega$$

Periodic (a_k & b_k are FS coeffs of $F(t)$ & $g(t)$):

$$\frac{1}{T_0} \int_{T_0} F(t) g^*(t) dt = \sum_{k=-\infty}^{\infty} a_k b_k^*$$

Useful because sometimes integrals we want to take match this form, but are much easier to compute in one domain compared to the other (i.e. integrating over a sinc vs. integrating over a rect).

Derivation: Start with multiplication property:

$$\mathcal{F}[f(t)g(t)] = \frac{1}{2\pi} F(j\omega) * G(j\omega) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\lambda) G(j(\omega-\lambda)) d\lambda$$

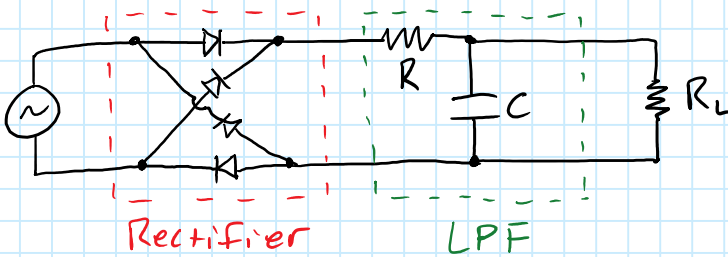
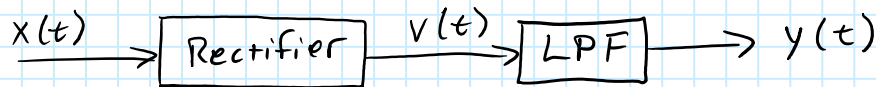
$$\mathcal{F}[f(t)g^*(t)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\lambda) G^*(j(\lambda-\omega)) d\lambda \quad \text{Conjugation Property}$$

$$\int_{-\infty}^{\infty} f(t)g^*(t) e^{-j\omega t} dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\lambda) G^*(j(\lambda-\omega)) d\lambda \quad \text{Definition of FT}$$

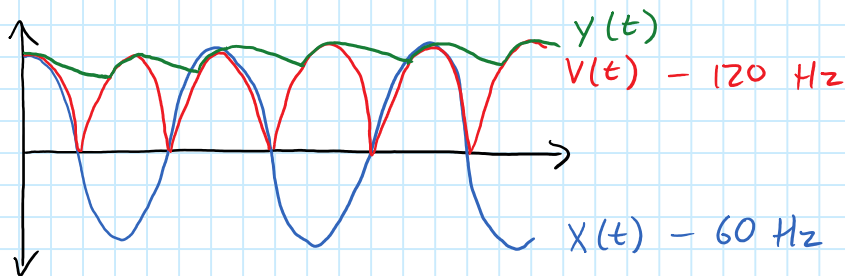
Evaluate at $\omega=0$

$$\begin{aligned} \int_{-\infty}^{\infty} f(t)g^*(t) dt &= \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\lambda) G^*(j\lambda) d\lambda \\ &= \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\omega) G^*(j\omega) d\omega \end{aligned} \quad \begin{array}{l} \text{Change } \lambda \text{ to } \\ \omega \text{ to get} \\ \text{Parseval's.} \end{array}$$

Example: Voltage Rectifier (AC to DC conversion / power supply)



Could have an inductor instead of a resistor in the LPF, but this is less common. These inductors can be big, heavy, expensive, and not very ideal (in behavior).



How much ripple is there in the output $y(t)$?

Quantify as $\frac{\text{RMS}_{\text{harmonics}}}{\text{DC value}}$

$\frac{\text{RMS amplitude of AC component}}{\text{RMS amplitude of DC}}$

Before the LPF:

FS coefficients of rectified cosine: $a_k = \frac{2}{\pi(1-4k^2)}$

Given by
Professor
Lanterman

$$f_0 = 120 \text{ Hz} \quad \omega_0 = 240\pi$$

$$T_0 = \frac{1}{120}$$

$$\text{RMS}^2 = \frac{1}{T_0} \int_{T_0} \left| \sum_{k \neq 0} a_k e^{-jk240\pi t} \right|^2 dt$$

FS synthesis sum for rectified cosine
minus the DC term ($k=0$).

$$a_0 = \frac{2}{\pi} = 0.6366 \quad \leftarrow \text{DC value}$$

$$a_1 = a_{-1} = \frac{2}{-3\pi} = -0.2122$$

$$a_2 = a_{-2} = \frac{2}{-15\pi} = -0.0424$$

$$a_3 = a_{-3} = \frac{2}{-35\pi} = -0.0182$$

And keep getting smaller...

AC Harmonics

Approx with 6 terms:

$$k = \pm 1, \pm 2, \pm 3$$

By Parseval's:

$$\text{RMS}^2 = \sum_{k \neq 0} |a_k|^2 \approx 2 \left((-0.2122)^2 + (-0.0424)^2 + (-0.0182)^2 \right)$$

$$\text{RMS}^2 \approx 0.0944$$

$$\text{RMS} \approx \sqrt{0.0944} = 0.3072 \quad \left(0.3001 \text{ from the fundamental} \right)$$

$$\% \text{ distortion} = \frac{0.3072}{0.6366} \times 100\% = 48\% \quad (\text{that's a lot})$$

After the LPF: Voltage divider: $H(j\omega) = \frac{\frac{1}{j\omega C}}{R + \frac{1}{j\omega C}} \cdot \frac{\left(\frac{j\omega}{R}\right)}{\left(\frac{j\omega}{R}\right)} = \frac{\frac{1}{RC}}{j\omega + \frac{1}{RC}}$

$$|H(j\omega)| = \left| \frac{1}{j\omega RC + 1} \right| = \frac{1}{\sqrt{(\omega RC)^2 + 1}}$$

$$y(t) = \sum_{k=-\infty}^{\infty} b_k e^{j240\pi k t} = \sum_{k=-\infty}^{\infty} a_k H(j240\pi k) e^{j240\pi k t}$$

Let $C = 1 \mu\text{F}$ and $R = 10 \text{ k}\Omega$

Evaluate $|H(j\omega)|$ for $k = \pm 0, 1, 2, 3$ & calc $|b_k|$

$$|H(j0)| = 1 \qquad |b_0| = (.6366)(1) = .6366$$

$$|H(j240\pi)| = 0.1315 \qquad |b_1| = (.2122)(.1315) = .0279$$

$$|H(j480\pi)| = 0.0662 \qquad |b_2| = (.0424)(.0662) = .0028$$

$$|H(j720\pi)| = 0.0442 \qquad |b_3| = (.0182)(.0442) = .0008$$

RMS ≈ 0.0412 (0.0395 from fundamental)

$$\% \text{ distortion} = \frac{0.0412}{0.6366} \times 100\% = 6.5\% \quad \text{much smaller}$$

May not be good enough for some applications (i.e. audio, where 120 Hz is very audible). Could design a better filter.